Sequential Multiuser Scheduling and Power Allocation for Clustered Cell-Free Massive MIMO Networks

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Abstract—Resource allocation is a fundamental task in cellfree (CF) massive multi-input multi-output (MIMO) systems, which can effectively improve the network performance. In this paper, we study the downlink of CF MIMO networks with network clustering and linear precoding, and develop a sequential multiuser scheduling algorithm based on greedy techniques and a gradient ascent (GA) power allocation algorithm for sum-rate maximization when imperfect channel state information (CSI) is considered. Numerical results show the superiority of the proposed sequential scheduling and power allocation scheme and algorithms to existing approaches while reducing the computational complexity and the signaling load.

Index Terms—Power allocation, user scheduling, massive MIMO, cell-free, clustering, complexity

I. INTRODUCTION

Cell-free (CF) massive MIMO networks introduced in [1] are distributed massive MIMO networks that include several access points (APs) over a geographic area to serve user equipments (UEs) in the same time and frequency resources, which provides uniform performance across users and improves coverage. Since CF networks in which all UEs are served by all APs, need to process all channels and signals in a central processing unit at the same time, they result in a huge burden to the processors and a substantial increase in costs. Therefore, it is necessary to use clustering techniques, which include network-centric and user-centric approaches [2] to reduce signaling and computational costs.

Resource allocation including power allocation and multiuser scheduling are key tasks for CF networks that can improve the system performance and have attracted a lot of attention in the literature [3]–[8]. Multiuser scheduling can reduce multiuser interference in CF networks, improving the system performance. In addition, if the number of receive antennas is larger than those of transmit antennas, it is impossible to support all the receivers which makes it necessary to employ multiuser scheduling. In this context, power allocation also leads to significant performance improvement in CF massive MIMO networks. In [9], the problem of multiuser scheduling, power allocation and beamforming in a usercentric cell-free MIMO wireless system has been solved by maximization of a weighted sum-rate (WSR) problem. The joint optimization of UE scheduling, power allocation and pilot length is investigated in [10] and the minimum ergodic user rate in the downlink transmission is maximized. In [11], the energy efficiency of a user-centric CF massive MIMO system is enhanced by solving a total grid power consumption minimization problem with a joint AP selection and user scheduling algorithm.

In this paper, we consider linear minimum mean square error (MMSE) and zero forcing (ZF) precoders and investigate the downlink of clustered CF massive MIMO networks with multiuser scheduling and power allocation. In particular, we develop a sequential multiuser scheduling and power allocation (SMSPA) scheme based on enhanced greedy and GA techniques to maximize the sum-rate. The proposed enhanced subset greedy (ESG) technique approaches the performance of the optimal exhaustive search method while significant computational cost can be saved. The proposed GA algorithm maximizes the sum-rate and is performed after scheduling the desired UEs set. Simulations show that the proposed SMSPA scheme, ESG and GA algorithms outperform competing approaches.

Notation: Throughout the paper, $\|.\|_F$ denotes the Frobenius norm, \mathbf{I}_N denotes the $N \times N$ identity matrix, the complex normal distribution is represented by $\mathcal{CN}(.,.)$, superscripts T , *, and H denote transpose, complex conjugate and hermitian operations respectively, $\mathcal{A} \cup \mathcal{B}$ is union of sets \mathcal{A} and \mathcal{B} , and $\mathcal{A} \setminus \mathcal{B}$ shows exclusion of set \mathcal{B} from set \mathcal{A} .

II. SYSTEM MODEL

The downlink of a CF massive MIMO network is considered where K uniformly distributed single-antenna UEs are supported by M single-antenna APs. Then, the CF network is clustered by dividing the network area into C non-overlapping clusters so that the cluster c includes K_c uniformly distributed single-antenna UEs and M_c single-antenna APs. We assume that the number of UEs is much larger than the number of APs so that K >> M for the CF network and $K_c >> M_c$ for cluster c of the clustered CF network, which requires the scheduling of a subset of $n_c \leq M_c$ UEs per cluster.

A. Cell-free massive MIMO network and the clustered context

In the CF network, the channel coefficient between AP mand UE u is shown by $g_{m,u} = \sqrt{\beta_{m,u}}h_{m,u}$, [1], including the large scale fading $\beta_{m,u}$ and the small-scale fading $h_{m,u} \sim C\mathcal{N}(0,1)$ defined as independent and identically distributed (i.i.d.) random variables (RVs) constant during a coherence interval and independent over different coherence intervals. After scheduling $n \leq M$ out of K UEs, the received signal is given by

$$\mathbf{y} = \sqrt{\rho_f} \mathbf{G}^T \mathbf{P} \mathbf{x} + \mathbf{w}$$

= $\sqrt{\rho_f} \hat{\mathbf{G}}^T \mathbf{P} \mathbf{x} + \sqrt{\rho_f} \tilde{\mathbf{G}}^T \mathbf{P} \mathbf{x} + \mathbf{w}$ (1)

where ρ_f is the maximum transmitted power of each antenna, $\mathbf{G} = \hat{\mathbf{G}} + \tilde{\mathbf{G}}$ is the $M \times n$ channel matrix, in which $\hat{\mathbf{G}}$ is the channel estimate, $\tilde{\mathbf{G}}$ is the estimation error that models the CSI imperfection and $[\mathbf{G}]_{m,u} = g_{m,u}, m \in \{1, \dots, M\},$ $u \in \{1, \dots, n\}, \mathbf{P}$ is the $M \times K$ linear precoder matrix such as MMSE or ZF, $\mathbf{x} = [x_1, \dots, x_n]^T$ is the zero mean symbol vector with mutually independent elements and independent of the channel coefficients $\mathbf{x} \sim C\mathcal{N}(\mathbf{0}, \mathbf{I}_n)$, and $\mathbf{w} = [w_1, \cdots, w_n]^T$ is the additive noise vector with $\mathbf{w} \sim C\mathcal{N}(0, \sigma_w^2 \mathbf{I}_n)$ and statistically independent of the signal vector. For Gaussian signaling, the sum-rate of the CF system is given by

$$R_{CF} = \log_2\left(\det\left[\mathbf{R} + \mathbf{I}_n\right]\right),\tag{2}$$

where the covariance matrix \mathbf{R} is expressed by

$$\mathbf{R} = \rho_f \hat{\mathbf{G}}^T \mathbf{P} \mathbf{P}^H \hat{\mathbf{G}}^* \left(\rho_f \tilde{\mathbf{G}}^T \mathbf{P} \mathbf{P}^H \tilde{\mathbf{G}}^* + \sigma_w^2 \mathbf{I}_n \right)^{-1}$$
(3)

In the clustered CF network, after scheduling $n_c \leq M_c$ out of K_c UEs, the received signal at cluster c is

$$\mathbf{y}_{c} = \sqrt{\rho_{f}} \mathbf{\hat{G}}_{cc}^{T} \mathbf{P}_{c} \mathbf{x}_{c} + \sqrt{\rho_{f}} \mathbf{\hat{G}}_{cc}^{T} \mathbf{P}_{c} \mathbf{x}_{c} + \sum_{i=1, i \neq c}^{C} \sqrt{\rho_{f}} \mathbf{\hat{G}}_{ic}^{T} \mathbf{P}_{i} \mathbf{x}_{i} + \sum_{i=1, i \neq c}^{C} \sqrt{\rho_{f}} \mathbf{\tilde{G}}_{ic}^{T} \mathbf{P}_{i} \mathbf{x}_{i} + \mathbf{w}_{c}$$
(4)

where $\mathbf{G}_{ic} = \hat{\mathbf{G}}_{ic} + \tilde{\mathbf{G}}_{ic}$ is $M_i \times n_c$ channel from APs of the cluster *i* to the UEs of the cluster *c*, \mathbf{P}_i is $M_i \times n_c$ linear precoding matrix, $\mathbf{x}_i = [x_{i1}, \dots, x_{inc}]^T$, $\mathbf{x}_i \sim \mathcal{CN}(\mathbf{0}, \mathbf{I}_{nc})$ is the symbol vector of the cluster *i*, $i \in \{1, 2, \dots, C\}$, and $\mathbf{w}_c = [w_{c_1}, \dots, w_{c_{nc}}]^T$ is the additive noise vector with $\mathbf{w}_c \sim \mathcal{CN}(\mathbf{0}, \sigma_w^2 \mathbf{I}_{nc})$. Therefore, the sum-rate of the cluster *c* in this network is given by

$$R_{c} = \log_{2} \left(\det \left[\left(\rho_{f} \hat{\mathbf{G}}_{cc}^{T} \mathbf{P}_{c} \mathbf{P}_{c}^{H} \hat{\mathbf{G}}_{cc}^{*} \right) \mathbf{R}_{c}^{-1} + \mathbf{I}_{n_{c}} \right] \right)$$
(5)

and the covariance matrix \mathbf{R}_c is described by

$$\mathbf{R}_{c} = E \left[\left(\mathbf{y}_{c} - \sqrt{\rho_{f}} \hat{\mathbf{G}}_{cc}^{T} \mathbf{P}_{c} \mathbf{x}_{c} \right) \left(\mathbf{y}_{c} - \sqrt{\rho_{f}} \hat{\mathbf{G}}_{cc}^{T} \mathbf{P}_{c} \mathbf{x}_{c} \right)^{H} \right]$$
$$= \rho_{f} \tilde{\mathbf{G}}_{cc}^{T} \mathbf{P}_{c} \mathbf{P}_{c}^{H} \tilde{\mathbf{G}}_{cc}^{*} + \sum_{i=1, i \neq c}^{C} \rho_{f} \hat{\mathbf{G}}_{ic}^{T} \mathbf{P}_{i} \mathbf{P}_{i}^{H} \hat{\mathbf{G}}_{ic}^{*}$$
$$+ \sum_{i=1, i \neq c}^{C} \rho_{f} \tilde{\mathbf{G}}_{ic}^{T} \mathbf{P}_{i} \mathbf{P}_{i}^{H} \tilde{\mathbf{G}}_{ic}^{*} + \sigma_{w}^{2} \mathbf{I}_{n_{c}}$$
(6)

where \mathbf{x}_c and \mathbf{w}_c are statistically independent. Finally, sumrate of the total clustered network is given by

$$R_{cl} = \sum_{c=1}^{C} R_c.$$
 (7)

III. PROPOSED SEQUENTIAL MULTIUSER SCHEDULING AND POWER ALLOCATION

In this section, we detail the proposed SMSPA scheme for multiuser scheduling and power allocation in clustered CF networks, which is outlined in Fig. 1. In particular, the SMSPA scheme employs an enhanced greedy algorithm for multiuser scheduling using the method presented in [12], in conjunction with a power allocation algorithm based on the GA method to maximize the sum-rate of the network. Unlike the enhanced greedy algorithm of the reference [13] which used equal power loading, we perform both scheduling and power allocation. Specifically, we first consider equal power loading and then employ the proposed greedy multiuser scheduling to schedule the best set using the sum-rate criterion and after that GA power allocation algorithm is employed.



Fig. 1. Block diagram of the proposed SMSPA resource allocation.

A. Proposed ESG multiuser scheduling algorithm

In cluster c including $\mathcal{U}_c = \{1, \ldots, K_c\}$ as UEs and M_c number of APs ($K_c > M_c$), in order to schedule a specific number of UEs such as n_c so that $n_c \leq M_c$, we first adapt a greedy method similar to the approach applied in [14] to select the first set of UEs. However, unlike [14] we apply the MMSE precoder instead of the ZF precoder so that we can obtain a better performance and refine the search algorithm. In addition, the number of selected UEs is prespecified. For the selected set S_{n_c} which results in a row-reduced channel matrix $\mathbf{G}_{cc}(S_{n_c})$, we aim at obtaining the solution to the optimization problem

$$\max_{S_{n_c}} R_{MMSE} \left(S_{n_c} \right)$$
subject to $\|\mathbf{P}_c \left(S_{n_c} \right) \|_{E}^{2} < P.$
(8)

where $R_{MMSE}(S_{n_c})$ is defined as the sum-rate with the MMSE precoder when S_{n_c} is the set of intended users, P is upper limit to the covariance matrix of the received signal, trace $[\mathbf{C_x}] \leq P$, and $\mathbf{P}_c(S_{n_c}) = \mathbf{W}_c \mathbf{D}_c$ is the precoding

matrix including the normalized MMSE weight matrix $\mathbf{W}_{c} \in$ $\mathbb{C}^{M_c \times n_c}$ and the power allocation matrix \mathbf{D}_c defined as

$$\mathbf{D}_{c} = \begin{bmatrix} \sqrt{p_{1}} & 0 & \cdots & 0\\ 0 & \sqrt{p_{2}} & \cdots & 0\\ \vdots & \vdots & \cdots & \vdots\\ 0 & 0 & \cdots & \sqrt{p_{n_{c}}} \end{bmatrix} = \operatorname{diag}(\mathbf{d}_{c})$$
$$, \mathbf{d}_{c} = \begin{bmatrix} \sqrt{p_{1}} & \sqrt{p_{2}} & \cdots & \sqrt{p_{n_{c}}} \end{bmatrix}^{T}. \quad (9)$$

We consider GA power allocation as described in Section III-B and the first set of UEs $S_{n_c(1)}$ is obtained using the first stage of Algorithm 1. In order to assess more sets of the users so that we can approach the optimal set achievable by the exhaustive search method, we consider $S_{n_c(2)}$ as the second set of UEs which is different from the set $S_{n_c(1)}$ in only one UE. To obtain $S_{n_c(2)}$, we remove the UE which has the lowest channel power among the UEs of the set $S_{n_c(1)}$ shown by u_r and replace it with the UE which possesses the most power among the UEs other than the set $S_{n_c(1)}$ shown by u_a . Thus, we can show the removed and added UEs as follows, respectively,

$$u_{r(1)} = \underset{u \in S_{n_c(1)}}{\operatorname{arg\,min}} \mathbf{g}_u^H \mathbf{g}_u \tag{10}$$

2

2

2 2

$$u_{a(1)} = \underset{u \in \mathcal{U}_{re(1)}}{\operatorname{arg\,max}} \mathbf{g}_{u}^{H} \mathbf{g}_{u}$$
(11)

where $\mathbf{g}_{u} = \begin{bmatrix} g_{1,u} & \cdots & g_{M_{c},u} \end{bmatrix}^{T}$ is the channel vector to UE u, and $\mathcal{U}_{re(1)} = \mathcal{U}_c \setminus S_{n_c(1)}$ shows the remaining UE set other than $S_{n_c(1)}$. The same process is done for the second set of UEs to find the third set. We continue to find new sets until we obtain $K_c - n_c$ sets of UEs beside the first set. We obtain the *i*th selected set of UEs and the *i*th remaining set of UEs as follows, respectively,

$$S_{n_c(i)} = \left(S_{n_c(i-1)} \setminus u_{r(i-1)}\right) \cup u_{a(i-1)}$$
(12)

$$\mathcal{U}_{re(i)} = \mathcal{U}_{re(i-1)} \setminus u_{a(i-1)} \tag{13}$$

where $i \in \{2, \dots, K_c - n_c + 1\}$. Finally, the desired set $S_{n_{cd}}$ among the obtained sets, is the set which results in the highest sum-rate $R_c(S_{n_c(J)}), J \in \{1, \dots, K_c - n_c + 1\}$ as derived in (5).

Algorithm 1: Proposed C-ESG Scheduling Algorithm.			
1 j=1 % first stage			
2	set $l = 1;$		
3	find a user such that		
4	$u_1 = \operatorname{argmax} \mathbf{g}_u^H \mathbf{g}_u;$		
	$u \in \mathcal{U}_c$		
5	set $U_1 = u_1$ and denote the achieved rate		
6	$R_{MMSE}\left(U_{1} ight);$		

B. GA power allocation algorithm

In Equation (4), the first part of the right hand side is the desired signal and remaining parts are the terms associated with imperfect CSI, inter-cluster interference and the noise. Therefore, by applying the power allocation, we rewrite the estimated received signal at the cluster c as follows

$$\mathbf{y}_{c} = \sqrt{\rho_{f}} \hat{\mathbf{G}}_{cc}^{T} \mathbf{W}_{c} \mathbf{D}_{c} \mathbf{x}_{c} + \sqrt{\rho_{f}} \tilde{\mathbf{G}}_{cc}^{T} \mathbf{P}_{c} \mathbf{x}_{c} + \sum_{i=1, i \neq c}^{C} \sqrt{\rho_{f}} \mathbf{G}_{ic}^{T} \mathbf{P}_{i} \mathbf{x}_{i} + \mathbf{w}_{c}$$
(14)

For sum-rate maximization, we combine the received signal of the UEs using a linear receiver $\mathbf{a} = \frac{1}{\sqrt{K_c}} \mathbf{1}_{K_c}^T$, where $\mathbf{1}_{K_c}$ is a $1 \times K_c$ vector of all 1 entities so that $\mathbf{a}^H \mathbf{a} = 1$ and we obtain a simpler expression for sum-rate [15]. After finding the power loading factors using the simplified sum-rate expression, we apply the obtained power allocation matrix in the sum-rate expressions defined in equations (2) and (5) to determine the sum-rates.

Thus, the combined received signal and the power ratio of the desired part of the signal to the interference and noise (SINR) are given as follows, respectively,

$$\mathbf{a}^{T}\mathbf{y}_{c} = \sqrt{\rho_{f}}\mathbf{a}^{T}\hat{\mathbf{G}}_{cc}^{T}\mathbf{W}_{c}\mathbf{D}_{c}\mathbf{x}_{c} + \sqrt{\rho_{f}}\mathbf{a}^{T}\tilde{\mathbf{G}}_{cc}^{T}\mathbf{P}_{c}\mathbf{x}_{c}$$
$$+ \sum_{i=1, i \neq c}^{C}\sqrt{\rho_{f}}\mathbf{a}^{T}\mathbf{G}_{ic}^{T}\mathbf{P}_{i}\mathbf{x}_{i} + \mathbf{a}^{T}\mathbf{w}_{c}$$
(15)

$$\text{SINR} = \frac{\rho_f}{\sigma_w^2} \frac{\mathbf{a}^T \hat{\mathbf{G}}_{cc}^T \mathbf{W}_c \mathbf{D}_c \mathbf{D}_c^H \mathbf{W}_c^H \hat{\mathbf{G}}_{cc}^* \mathbf{a}}{\mathbf{a}^T \mathbf{Z} \mathbf{a}}$$
(16)

where

$$\mathbf{Z} = \tilde{\mathbf{G}}_{cc}^{T} \mathbf{P}_{c} \mathbf{P}_{c}^{H} \tilde{\mathbf{G}}_{cc}^{*} + \sum_{i=1, i \neq c}^{C} \rho_{f} \mathbf{G}_{ic}^{T} \mathbf{P}_{i} \mathbf{P}_{i}^{H} \mathbf{G}_{ic}^{T} + \mathbf{I} \quad (17)$$

Assuming Gaussian signaling, the rate expression is obtained by $\frac{1}{2} \log_2 (1 + \text{SINR})$. Accordingly, the sum-rate expression is given by

$$SR = \frac{1}{2}\log_2\left[1 + \frac{\rho_f}{\sigma_w^2} \frac{\mathbf{a}^T \hat{\mathbf{G}}_{cc}^T \mathbf{W}_c \mathbf{D}_c \mathbf{D}_c^H \mathbf{W}_c^H \hat{\mathbf{G}}_{cc}^* \mathbf{a}}{\mathbf{a}^T \mathbf{Z} \mathbf{a}}\right].$$
(18)

Equation (18) is similar to $\frac{1}{2}\log_2(1+bx)$ where $b = \frac{\rho_f}{\sigma_w^2 \mathbf{a}^T \mathbf{Z} \mathbf{a}}$ and $x = \mathbf{a}^T \hat{\mathbf{G}}_{cc}^T \mathbf{W}_c \mathbf{D}_c \mathbf{D}_c^H \mathbf{W}_c^H \hat{\mathbf{G}}_{cc}^* \mathbf{a}$ which is a monotonically increasing function of x, b > 0. Thus, we can maximize x which is equivalent to the sum-rate using the following problem

$$\max_{\mathbf{d}_{c}} \left(\mathbf{a}^{T} \hat{\mathbf{G}}_{cc}^{T} \mathbf{W}_{c} \operatorname{diag}\left(\mathbf{d}_{c}\right) \operatorname{diag}\left(\mathbf{d}_{c}\right)^{H} \mathbf{W}_{c}^{H} \hat{\mathbf{G}}_{cc}^{*} \mathbf{a} \right)$$
subject to $\left\| \mathbf{W}_{c} \operatorname{diag}\left(\mathbf{d}_{c}\right) \right\|^{2} \leq P.$
(19)

Since the objective function x is scalar, trace(x) = x. Therefore, by taking the derivative of the objective function with respect to the power loading matrix \mathbf{D}_c and using the equality $\frac{\partial \operatorname{trace}(\mathbf{AB})}{\partial \mathbf{A}} = \mathbf{B} \odot \mathbf{I}$ where \mathbf{A} is a diagonal matrix and \odot shows the Hadamard product, we obtain

$$\frac{\partial x}{\partial \mathbf{D}_{c}} = 2 \left(\mathbf{W}_{c}^{H} \hat{\mathbf{G}}_{cc}^{*} \mathbf{a} \mathbf{a}^{T} \hat{\mathbf{G}}_{cc}^{T} \mathbf{W}_{c} \operatorname{diag}\left(\mathbf{d}_{c}\right) \right) \odot \mathbf{I}$$
(20)

We can use a stochastic GA approach to update the power allocation coefficients as follows

$$\mathbf{d}_{c}(i) = \mathbf{d}_{c}(i-1) + \lambda \frac{\partial x}{\partial \mathbf{D}_{c}} = \mathbf{d}_{c}(i-1) + 2\lambda \left(\mathbf{W}_{c}^{H} \hat{\mathbf{G}}_{cc}^{*} \mathbf{a} \mathbf{a}^{T} \hat{\mathbf{G}}_{cc}^{T} \mathbf{W}_{c} \operatorname{diag}\left(\mathbf{d}_{c}(i-1)\right) \right) \odot \mathbf{I}$$
(21)

where *i* and λ represent the iteration index and the positive step size, respectively. Before running the adaptive algorithm, the transmit power constraint should be satisfied so that $\|\mathbf{W}_c \operatorname{diag}(\mathbf{d}_c)\|_F^2 = \|\mathbf{P}_c\|_F^2 \leq P$. Therefore, the power scaling factor $\eta = \sqrt{\frac{\operatorname{trace}(\mathbf{P}_c \mathbf{P}_c^H)}{\operatorname{trace}(\mathbf{W}_c \operatorname{diag}(\mathbf{d}_c, \mathbf{d}_c) \mathbf{W}_c^H)}}$ is employed in each iteration to scale the coefficients properly. The adaptive power allocation is summarized in Algorithm 2 where \mathbf{I}_t iterations are used.

Algorithm 2: GA Power Allocation Algorithm for Sum-Rate Maximization.

$$\begin{array}{c|c} \mathbf{I} \text{ Input } \mathbf{G}_{cc}, \mathbf{P}_{c}, \mathbf{W}_{c}, \text{ a and } \lambda \\ \mathbf{2} \ \mathbf{d}_{c} (1) = \mathbf{0} \\ \mathbf{3} \ \mathbf{for} \ i = 2 \ \mathrm{to} \ \mathbf{I}_{t} \ \mathbf{do} \\ \mathbf{4} & \begin{vmatrix} \frac{\partial x}{\partial \mathbf{D}_{c}} = \\ 2 \left(\mathbf{W}_{c}^{H} \hat{\mathbf{G}}_{cc}^{*} \mathbf{aa}^{T} \hat{\mathbf{G}}_{cc}^{T} \mathbf{W}_{c} \mathrm{diag} \left(\mathbf{d}_{c} \left(i - 1 \right) \right) \right) \odot \mathbf{I}; \\ \mathbf{5} & \mathbf{d}_{c} \left(i \right) = \mathbf{d}_{c} \left(i - 1 \right) + \lambda \frac{\partial x(\varepsilon)}{\partial \mathbf{D}_{c}}; \\ \mathbf{6} & \text{if trace} \left(\mathbf{W}_{c} \mathrm{diag} \left(\mathbf{d}_{c} \left(i \right) \cdot \mathbf{d}_{c} \left(i \right) \right) \mathbf{W}_{c}^{H} \right) \neq \\ \mathrm{trace} \left(\mathbf{P}_{c} \mathbf{P}_{c}^{H} \right) \ \mathbf{then} \\ \mathbf{7} & \begin{vmatrix} \eta = \sqrt{\frac{\mathrm{trace}(\mathbf{P}_{c} \mathbf{P}_{c}^{H})} \\ \mathbf{d}_{c} \left(i \right) = \eta \mathbf{d}_{c} \left(i \right); \\ \mathbf{9} & \end{vmatrix} \begin{array}{c} \mathbf{end} \\ \mathbf{no \ end} \end{array}$$

IV. SIMULATIONS

In order to assess the proposed SMSPA resource allocation scheme, we compare the sum-rate of the networks that use the proposed ESG, standard greedy (SG), exhaustive search (ES) or WSR user scheduling techniques and the proposed GA power allocation or equal power loading (EPL). Note that we have adapted the WSR technique proposed in [9] to the clustering method we have implemented so that the WSR of the UEs in each cluster supported by the corresponding APs is maximized. The considered CF network is a squared area with the side length of 400 m equipped with M randomly located APs and K uniformly distributed UEs. Applying network-centric clustering, we have considered C = 4 non-overlapping clusters, where cluster c includes M_c randomly located APs and K_c uniformly distributed UEs and the power allocation is performed using GA algorithm. The large scale coefficient in CF channel coefficient is modeled as $\beta_{m,k} = \text{PL}_{m,k} \times 10^{\frac{\sigma_{sh}z_{m,k}}{10}}$ where $10^{\frac{\sigma_{sh}z_{m,k}}{10}}$ is the shadow fading with $\sigma_{sh} = 8\text{dB}$, $z_{m,k} \sim \mathcal{N}(0,1)$, and $\text{PL}_{m,k}$ is the path loss modeled as [16]

$$PL_{m,k} = \begin{cases} -D - 35 \log_{10} (d_{m,k}), \text{ if } d_{m,k} > d_1 \\ -D - 10 \log_{10} \left(d_1^{1.5} d_{m,k}^2 \right), \text{ if } d_0 < d_{m,k} \le d_1 \\ -D - 10 \log_{10} \left(d_1^{1.5} d_0^2 \right), \text{ if } d_{m,k} \le d_0 \end{cases}$$
(22)

where $d_{m,k}$ is the distance between the *m*th AP and *k*th UE and D is

$$D = 46.3 + 33.9 \log_{10} (f) - 13.82 \log_{10} (h_{AP}) - [1.11 \log_{10} (f) - 0.7] h_u + 1.56 \log_{10} (f) - 0.8$$
(23)

where f = 1900MHz is the carrier frequency, $h_{AP} = 15$ m, $h_u = 1.5$ m are the AP and UE antenna heights, respectively, $d_0 = 10$ m and $d_1 = 50$ m. If $d_{m,k} \le d_1$ there is no shadowing.

In Fig. 2(a), the sum-rate performances of the proposed SMSPA scheme is assessed with the ESG scheduling algorithm using EPL or the GA power allocation when ZF or MMSE precoders are applied. While the sum-rates are increasing with the increase in the signal-to-noise ratio (SNR), the MMSE precoder outperforms the ZF precoder. In addition, the GA

power allocation yields significant performance improvement at low-to-medium SNR values.

Fig. 2(b) shows a comparison of different resource allocation techniques when the MMSE precoder is used. We employ a network with a small number of UEs while half of the UEs are scheduled so that we can show the results for the ES method as well as other methods. We notice that the proposed SMSPA resource allocation which has used the ESG and GA algorithms has outperformed other approaches and in the CF network the performance is close to that of the optimal ES method. As expected and according to Equation (4), CF shows better performance than that of the CLCF network because of the extra interference terms caused by other clusters. We clarify that Fig. 2 is plotted according to the sum-rate expressions of equations (2) and (5) and simplified sum-rate equation of (18) is used only to derive the power loading factors. However, as shown in Table I, the computational cost of the proposed SMSPA scheme and the signaling load as the number of channel parameters for CLCF network are substantially lower than CF.



Fig. 2. Performance of resource allocation schemes, (a): Comparison of the proposed SMSPA technique in CF networks for ZF and MMSE precoders with the system which has implemented the proposed ESG algorithm and equal power loading (EPL) (M = 64, K = 128, n = 24), (b): Comparison of the different resource allocation techniques in CF and CLCF networks adapting GA power allocation when MMSE precoder is used (M = 64, K = 16, n = 8).

TABLE I

Computational complexity of the proposed SMSPA scheme in floating point operations and the signaling load in parameters for CF and CLCF networks when M = 64, K = 128 and n = 64.

Network	CF	CLCF
Signaling load	24576	6144
Computational cost	1.3632×10^9	69.354×10^{6}

V. CONCLUSIONS

This work has investigated resource allocation and sum-rate performance of the CF and the clustered CF networks with ZF and MMSE precoders. An SMSPA resource allocation scheme is developed that is based on ESG multiuser scheduling and GA power allocation algorithms. Simulations have shown that the proposed SMSPA scheme has outperformed the existing methods and using the PA algorithm has also considerably improved the network performance compared with the EPL case. Additionally, in the case of the network clustering, a substantial computational complexity is saved using the proposed SMSPA scheme and the signaling load is much lower.

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